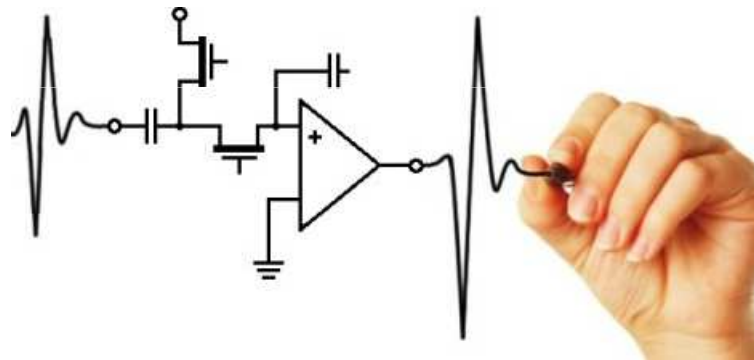


Semi-Empirical Optimization of CMOS Low-noise Amplifier for Biomedical Data Acquisition



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 - Orthogonal-Convex Optimization
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Prologue

- Design of neural recording interfaces.
[for biomedical measurements]
- Generic structure of neural recording interfaces.
- Low-Noise Amplifier (LNA): A Key Element.
- Design challenge.
[stringent noise-power-area trade-off]
- Application domain of biomedical LNAs.
[ECG, EEG, EMG]



Amplifier for Biomedical Data Acquisition

- Capacitive feedback network (CFN) topology.
- Prerequisites.
 - low power, low input-referred noise and high CMRR
- ♥ of the CFN LNA: Preamplifier.

Table 1

Signal	Frequency (Hz)	Amplitude (V)
ECG	0.05 – 250	$5 \times 10^{-6} - 8 \times 10^{-3}$
EEG	0.5 – 200	$2 \times 10^{-6} - 200 \times 10^{-6}$
EMG	0.01 – 10×10^3	$50 \times 10^{-6} - 10 \times 10^{-3}$

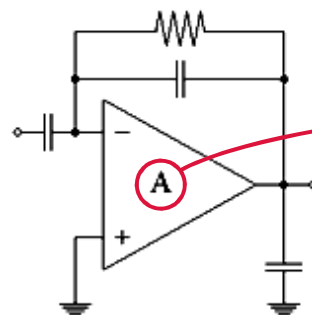


Figure 1

Transfer function

$$\approx \frac{\text{open-loop voltage gain}}{\left[1 + \frac{s}{\text{dominant pole}}\right] \left[1 + \frac{s}{\text{output pole}}\right] \left[1 + \frac{s}{\text{compensation pole}}\right]}$$

Figure 2

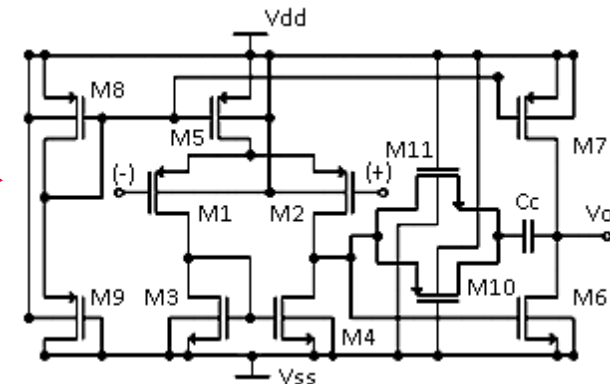
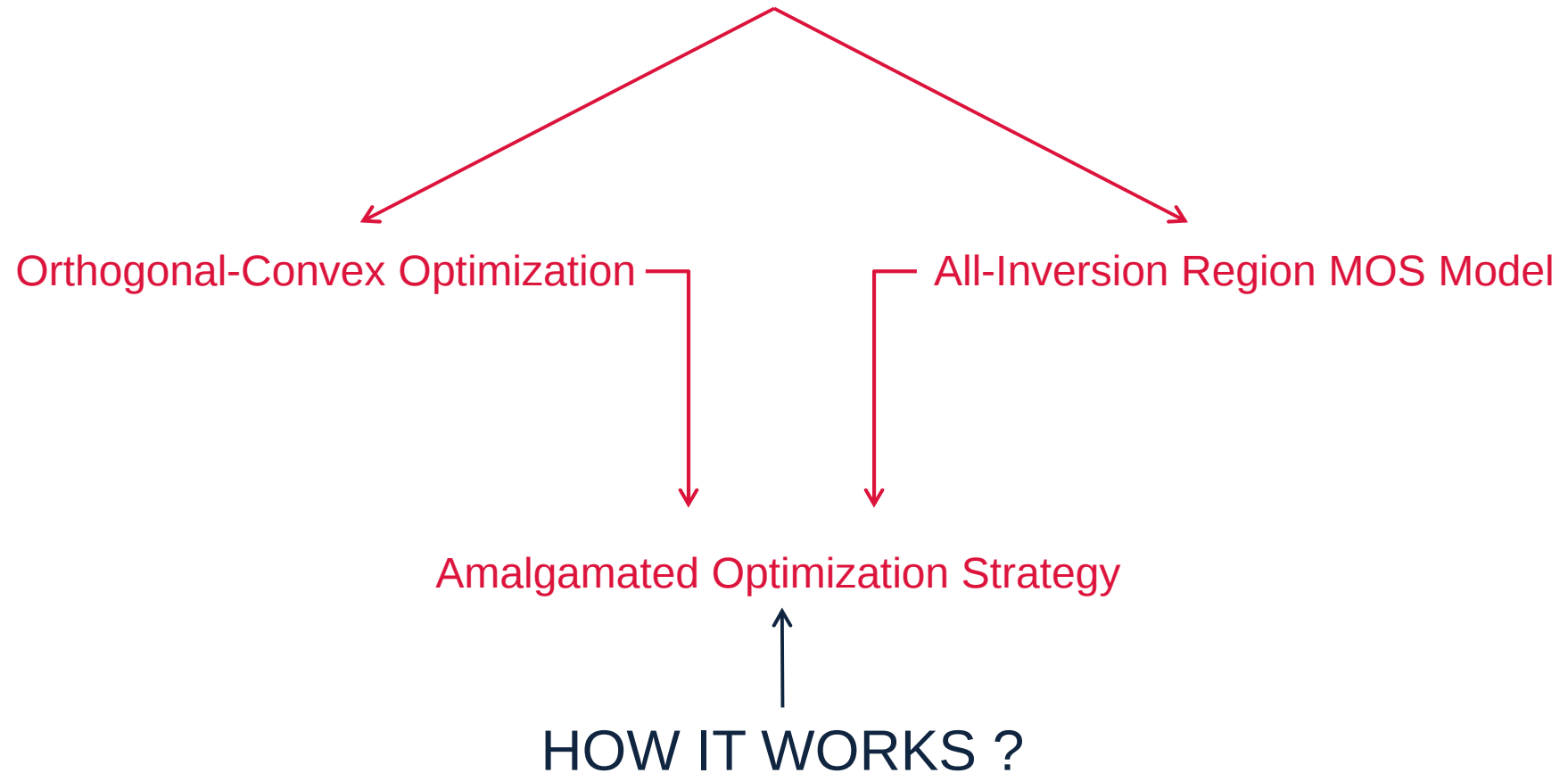


Figure 3

Optimization Methodology



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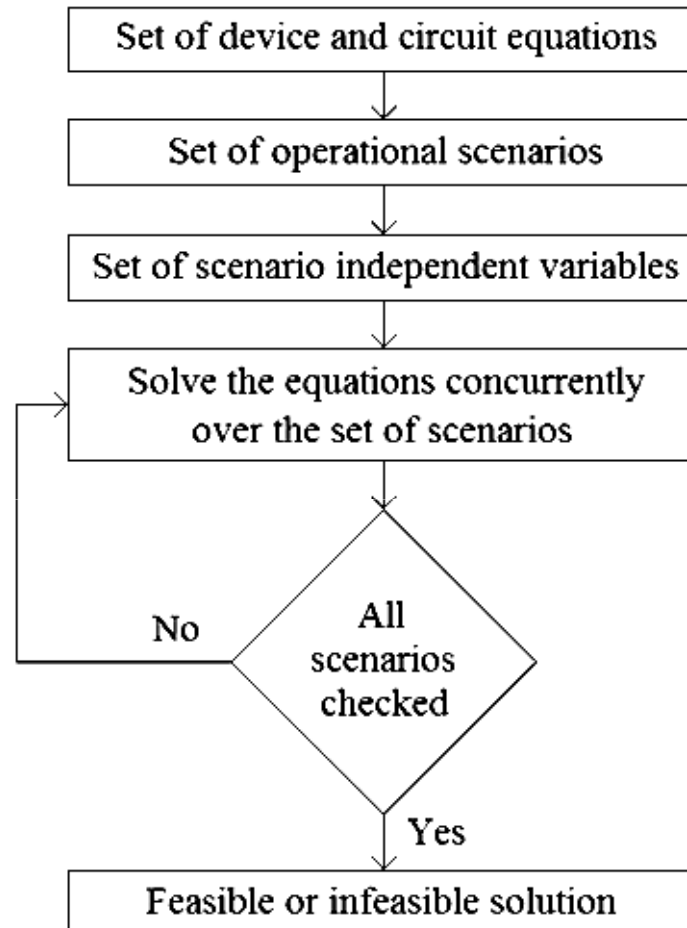


Figure 4

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Orthogonal-Convex Optimization

- Objective & Constraint Based Optimization.
- Problem Formulation:

$$\begin{aligned} &\text{optimize } f_0(x) \\ &\text{subject to } f_i(x) \leq 1, \quad i = (1, \dots, m) \\ &\quad \quad \quad g_i(x) = 1, \quad i = (1, \dots, p) \\ &\quad \quad \quad x_i > 0, \quad i = (1, \dots, n) \end{aligned}$$

- Monomial & Posynomial:

$$g(x) = cx_1^{a_1} \dots x_n^{a_n} \quad f(x) = \sum_{k=1}^K c_k g_k(x)$$

- Logarithmic Transformation of Variables.

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All-Inversion Region MOS Model

- Inversion Coefficient of MOSFET Optimization.

$$ICF = \frac{I_D}{I_0 S}$$

$$I_0 = 2n\mu C_{OX} \phi_T^2$$

Weak Inversion: $ICF < 0.1$

Strong Inversion: $ICF > 10$

Moderate Inversion: $0.1 < ICF < 10$

- Transconductance Efficiency.

$$\frac{g_m}{I_D} = \frac{1}{n\phi_T} \frac{2}{(1 + \sqrt{ICF + 1})}$$



Design Formulation

- Objective: Minimize Power Dissipation
- Constraints:
 - gain-bandwidth, voltage gain, common-mode rejection ratio, slew-rate, phase margin, noise spectral density
- Upper and/or Lower Bounds on Design Parameters.
- Example 1:

Open-loop Gain = Gain of 1st Stage x Gain of 2nd Stage

$$\begin{aligned} &= \left(\frac{g_{m2}}{g_{o2} + g_{o4}} \right) \left(\frac{g_{m6}}{g_{o6} + g_{o7}} \right) \\ &= \frac{\left(\frac{g_{m2}}{I_{D2}} I_{D2} \right)}{\left(\frac{I_{D2}}{V_{A2}} + \frac{I_{D4}}{V_{A4}} \right)} \frac{\left(\frac{g_{m6}}{I_{D6}} I_{D6} \right)}{\left(\frac{I_{D6}}{V_{A6}} + \frac{I_{D7}}{V_{A7}} \right)} \end{aligned}$$

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■ Example 2:

$$\begin{aligned}
 \text{Nulling Active Resistor} &= R_z \\
 &= \frac{1}{g_{m6}} \\
 &= \frac{1}{\frac{g_{m6}}{I_{D6}} I_{D6}}
 \end{aligned}$$

Table 2

■ Design Summary:

[180-nm BSIM3 CMOS Model]

Transistor length	$0.18\mu\text{m} \leq L \leq 1.8\mu\text{m}$
Device width	$0.18\mu\text{m} \leq W \leq 180.0\mu\text{m}$
Supply voltage	1.25V
Load capacitance	1pF
Oxide thickness	4.1nm
Open-loop voltage gain	$\geq 60\text{dB}$
Common-mode rejection ratio	$\geq 60\text{dB}$
Gain-bandwidth	$\geq 1\text{MHz}$
Slew-rate	$\geq 1\text{V}/\mu\text{S}$
Phase margin	$\geq 60^\circ$
Input-referred noise	$\leq 450\text{nV}/\sqrt{\text{Hz}}$
Power	Minimize



Result Interpretation

■ 2-Tier Process: Estimation and Verification.

Table 3

Transistor length	$L_1 = L_2$	1.8 μ m
	$L_3 = L_4$	1.8 μ m
	$L_5 = L_7 = L_8$	1.43658 μ m
	L_6	1.8 μ m
Transistor width	$W_1 = W_2$	180 μ m
	$W_3 = W_4$	180 μ m
	W_5	180 μ m
	W_6	20.87414 μ m
	W_7	10.43707 μ m
	W_8	1.43658 μ m
Nulling active resistance	R_Z	420.7278k Ω
Compensation capacitance	C_C	1.780275pF
Open-loop voltage gain	G	80.0dB
Common-mode rejection ratio	$CMRR$	82.70868dB
Gain-bandwidth	GBW	22.44995MHz
Slew-rate	SR	1V/ μ S
Phase margin	PM	78.07401°
Input-referred noise	$S_{in}(f)$	319.6656nV/ $\sqrt$ Hz
Power	P	64.8477 μ W

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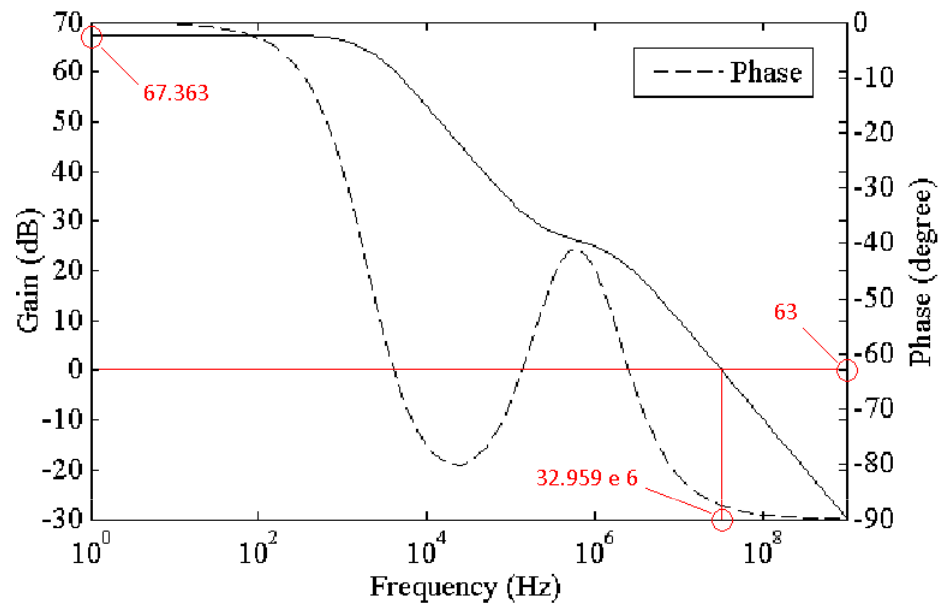


Figure 5

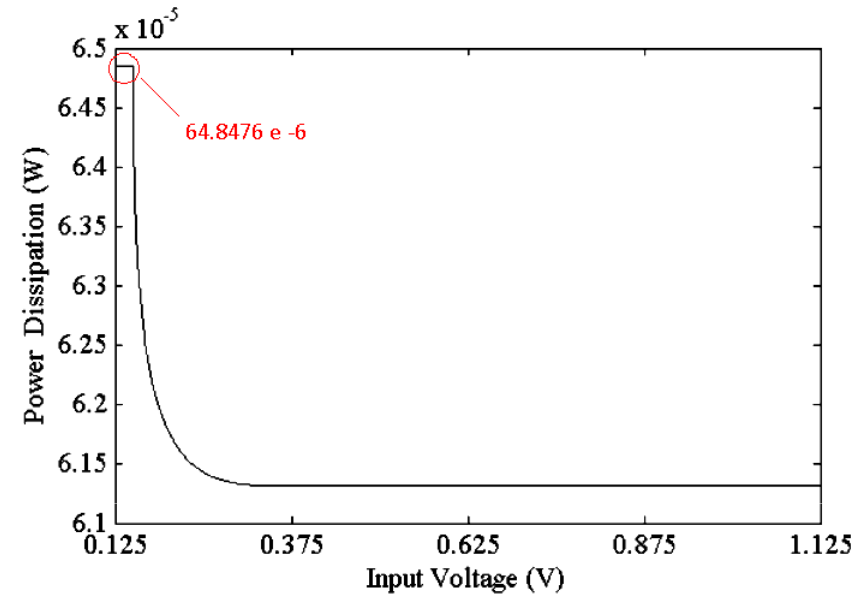


Figure 6

Table 4

Parameter	This method	SPICE	Unit
<i>G</i>	80.0	67.363	dB
<i>CMRR</i>	82.70868	94.124	dB
<i>GBW</i>	22.44995	32.959	MHz
<i>SR</i>	1	1.45	V/ μ S
<i>PM</i>	78.07401	63	degree
<i>P</i>	64.8477	64.8476	μ W

Epilogue

- Amalgamated Optimization Strategy:

effectiveness of Orthogonal-Convex Optimization

+

flexibility of All-Inversion MOSFET Model

- Pertains Prompt & Globally Optimal Solutions. 👍
- Increased Fidelity and Flexibility of Design. 👍
- Rigorous Initial Computation. 👉
- Limited Function Handling Capability. 👉



"Further Accuracy is Achievable by Adopting Judicious Design Strategies"



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QUESTIONS / SUGGESTIONS

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