



In this Lecture...

Frequency domain analysis

Discrete-Time Fourier Transform

Discrete Fourier Series or Discrete Fourier Transform

Introduction

- In most cases we want to know the frequency content of our signal
 - ▣ Why?
- Most popular analysis in frequency domain is based on work of Joseph Fourier
- We will cover two methods of frequency domain analysis of discrete-time signals
 - ▣ Discrete-Time Fourier Transform
 - ▣ Discrete Fourier Series or Discrete Fourier Transform

Spectrum of a signal

Fourier		Spectrum	
		Continuous	Discrete
Signal	Continuous	Fouriertransform	Fourierseries
	Discrete	Discrete-time Fouriertransform	DiscreteFourier transform Or DiscreteFourier series

Aperiodic Signal

Periodic Signal

Discrete Time Fourier Transform

- Applies to a-periodic discrete-time signal
- Mathematically calculated as,

$$X(\omega) = DTFT\{x[n]\} = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \quad -\pi < \omega < \pi$$

- The Inverse DTFT is calculated as

$$x[n] = IDTFT\{X(\omega)\} = \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega \quad -\infty < n < \infty$$

Periodicity?

- The Fourier transform of a Discrete-time signal

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$X(\omega + 2\pi k) = \sum_{n=-\infty}^{\infty} x[n] e^{-j(\omega + 2\pi k)n}$$

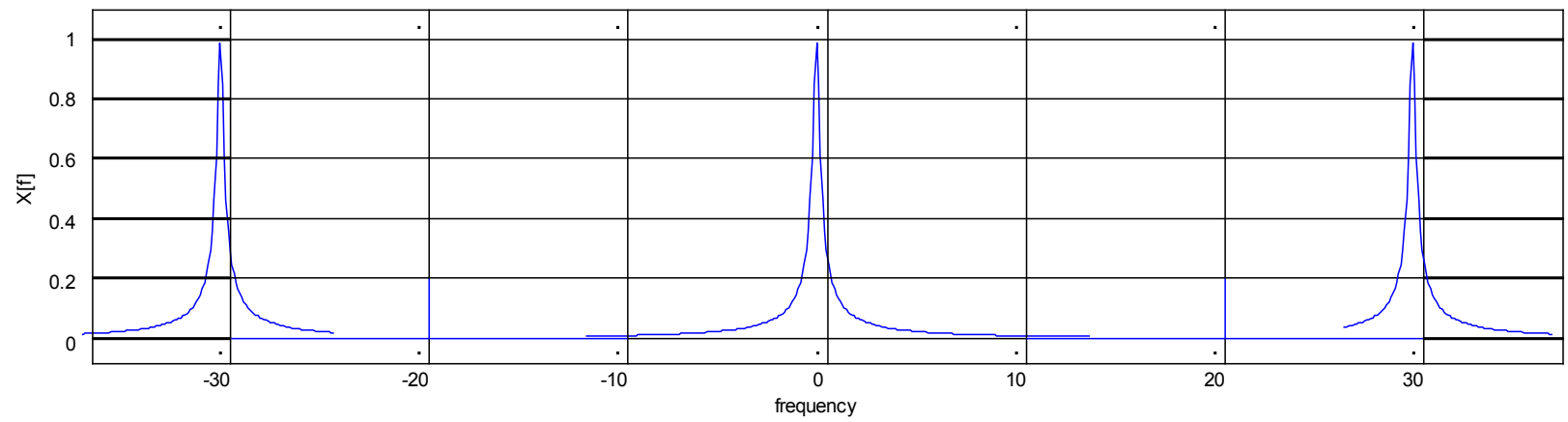
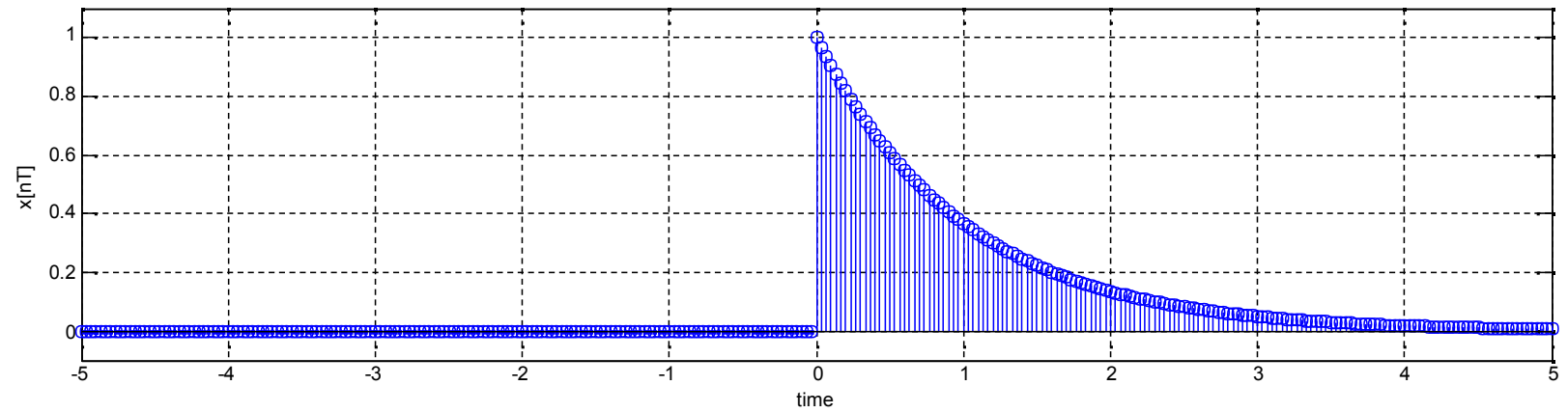
$$X(\omega + 2\pi k) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} e^{-j2\pi kn}$$

$$e^{-j2\pi kn}$$

$$e^{-j2\pi kn} = \cos(2\pi kn) - j \sin(2\pi kn) = 1$$

$$X(\omega + 2\pi k) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

$$X(\omega + 2\pi k) = X(\omega)$$



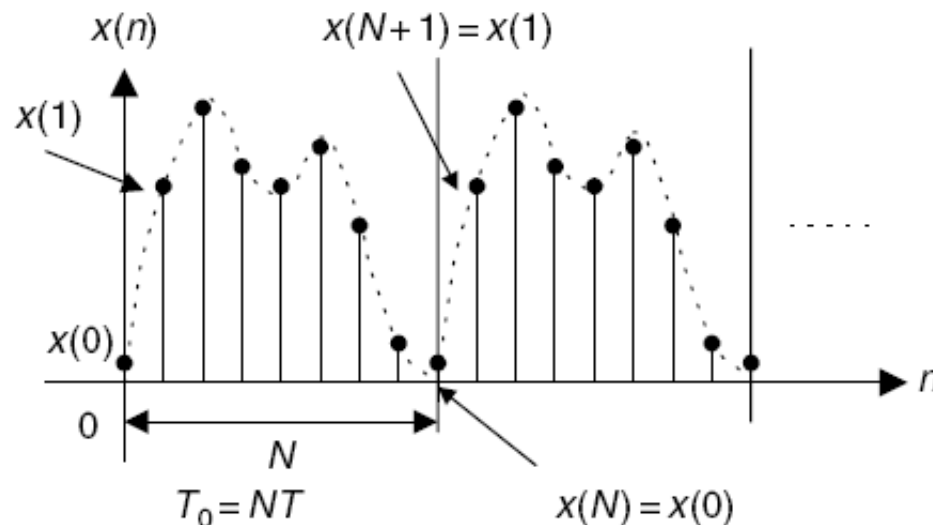
Points to ponder...

- The signal is discrete in time domain but continuous in frequency domain
 - ▣ The integration sign gives it away!
- Usually used for signals with infinite period
- DTFT gives exact frequency concentration for $w(\omega)$
 - ▣ But requires sequence to last $-\infty < n < \infty$
 - ▣ Requires infinite memory
- IDTFT also requires a infinite amount of memory to perform integration $-\pi < \omega < \pi$ and infinite processing time!

- In practice, we ***don't*** have infinite memory, we ***can't*** afford to have infinite processing time and the signal we want to analyze ***does not*** have infinite duration
- Besides, spectrum usually changes with time!
 - ▣ Music... changing notes
- Question for you
 - ▣ DTFT/IDTFT fails?
- DTFT/IDTFT is only good for simple analytical expressions like $u[n]$

Discrete Fourier Series

- Applies to periodic discrete-time signal
- Explanation by example!
- Suppose one wants to estimate the spectrum of discrete-time periodic signal



$x(t)$ is sampled at f_s Sampling
period $T = 1/f_s$

Fundamental period $T_0 = NT$

Fourier series for analog signal is given as

$$c_k = \frac{1}{T_0} \int_{T_0} x(t) e^{-jk\omega_0 t} dt \quad -\infty < k < \infty$$

Substituting $T_0 = NT$ and $\omega_0 = 2\pi/T_0$ and approximating the integration over one period by summation over one period, also t by nT , we get

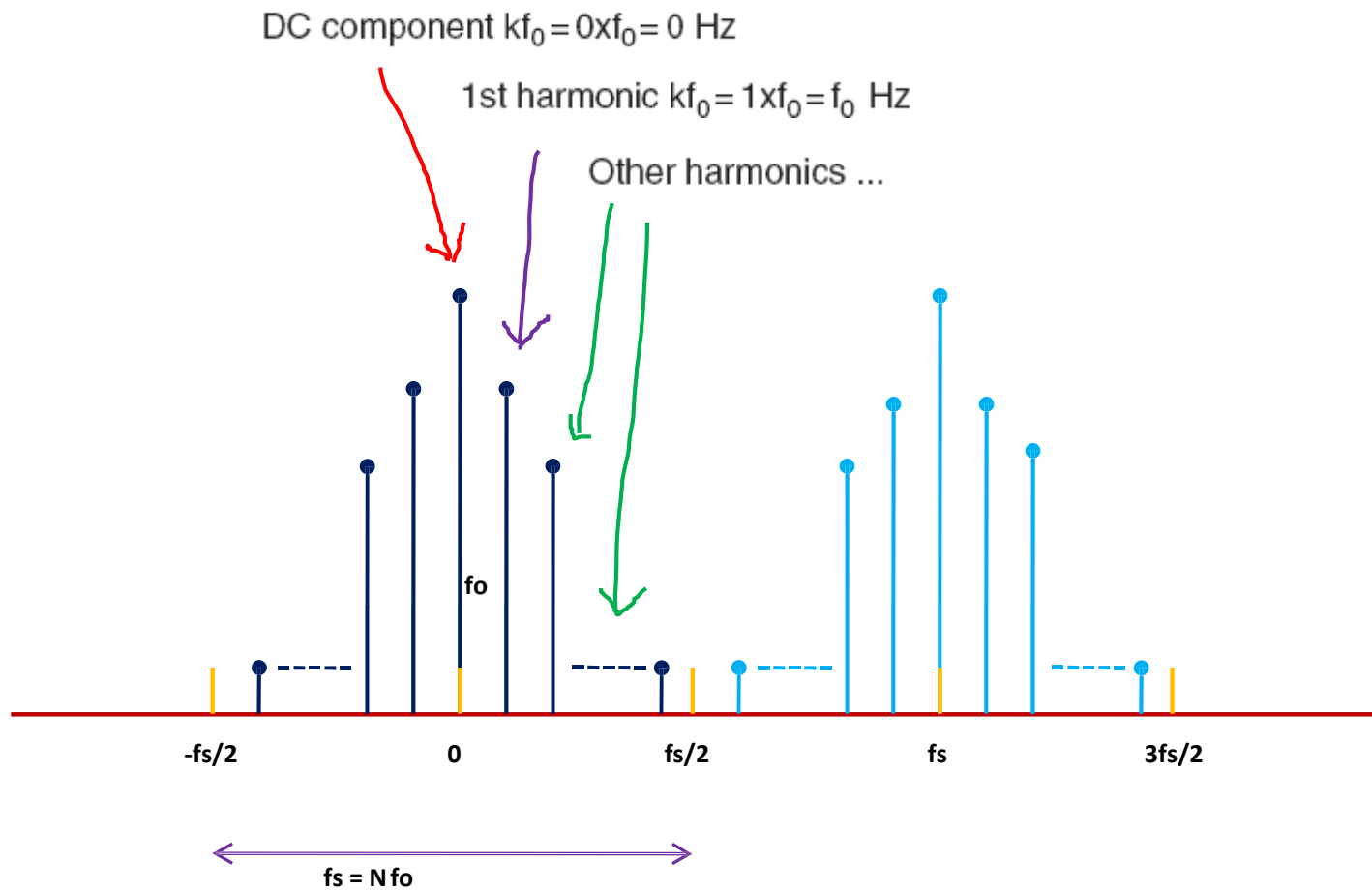
$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}} \quad -\infty < k < \infty$$

Analysis

$$\tilde{x}[n] = \sum_{k=0}^{N-1} c_k e^{j\frac{2\pi kn}{N}}$$

Synthesis

It can be easily proved that c_k is periodic with N (just replace k by $N+K$)
So that, $c_k = c_{k+N}$



Amplitude spectrum for c_k

Points to ponder...

- As displayed in Figure, only the line spectral portion between the frequency $-f_s/2$ and frequency $f_s/2$ (folding frequency) represents the frequency information of the periodic signal
- The spectral portion from $f_s/2$ to f_s is a copy of the spectrum in the negative frequency range from $-f_s/2$ to 0 Hz due to the spectrum being periodic for every Nf_0 Hz

- Amplitude spectral components from $f_s/2$ to f_s are really the folded version of the amplitude spectral components from 0 to $f_s/2$ in terms of $f_s - f$
- So we usually only calculate for coeffs for positive frequencies and use periodic property to find coeffs for negative frequency

i.e.

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi kn}{N}} \quad k = 0, 1, \dots, N-1$$

- For the k th harmonic the frequency is $f = k f_0$ Hz

Example 1: The periodic signal $x(t) = \sin(2\pi)t$, is sampled at 4 Hz

(a) Compute the spectrum c_k

(b) Plot the two sided spectrum

From the analog signal we can find the fundamental frequency

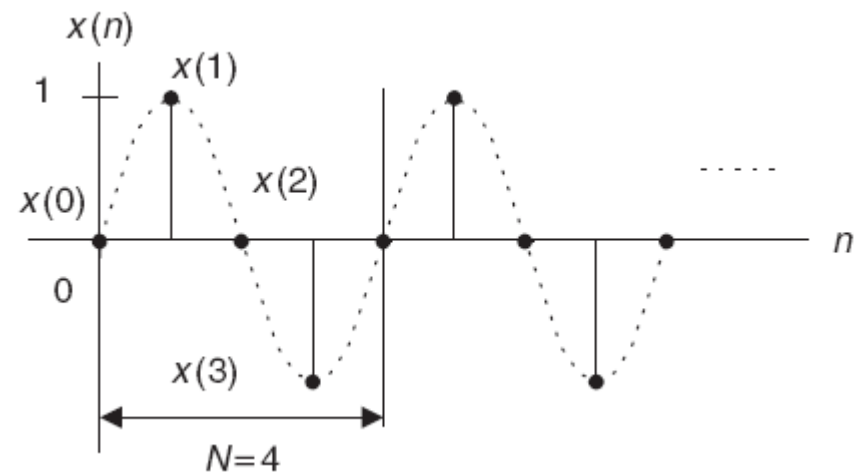
$$\omega_o = 2\pi \text{ rad/sec} \Rightarrow f_o = \frac{\omega_o}{2\pi} = \frac{2\pi}{2\pi} = 1 \text{ Hz}$$

And the fundamental period is $T_o = 1 \text{ sec}$

Using the sampling interval $T = \frac{1}{f_s} = \frac{1}{4} = 0.25 \text{ sec}$, we get the

Sampled signal as

$$x[n] = \sin(2\pi nT) = \sin(0.5\pi n)$$



Choosing the duration of one period we take the values of $x[n]$

$$x[0] = 0, \quad x[1] = 1, \quad x[2] = 0, \quad x[3] = -1$$

Solving the equation for DFS

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi kn}{N}}$$

$$c_0 = \frac{1}{4} \sum_{n=0}^3 x[n] = \frac{1}{4} \{x[0] + x[1] + x[2] + x[3]\} = \frac{1}{4} \{0 + 1 + 0 - 1\} = 0$$

$$c_1 = \frac{1}{4} \sum_{n=0}^3 x[n] e^{-j 2\pi k \frac{n}{4}} = \frac{1}{4} \{x[0] + x[1]e^{-j\pi/2} + x[2]e^{-j\pi} + x[3]e^{-j3\pi/2}\}$$

$$= \frac{1}{4} \{x[0] + x[1](-j) + x[2](-1) + x[3](j)\}$$

Using Euler's expansion
of $e^x = \cos(x) + j\sin(x)$

$$= \frac{1}{4} \{0 + 1(-j) + 0(-1) - 1(j)\} = \frac{1}{4} \{-2j\} = -j0.5$$

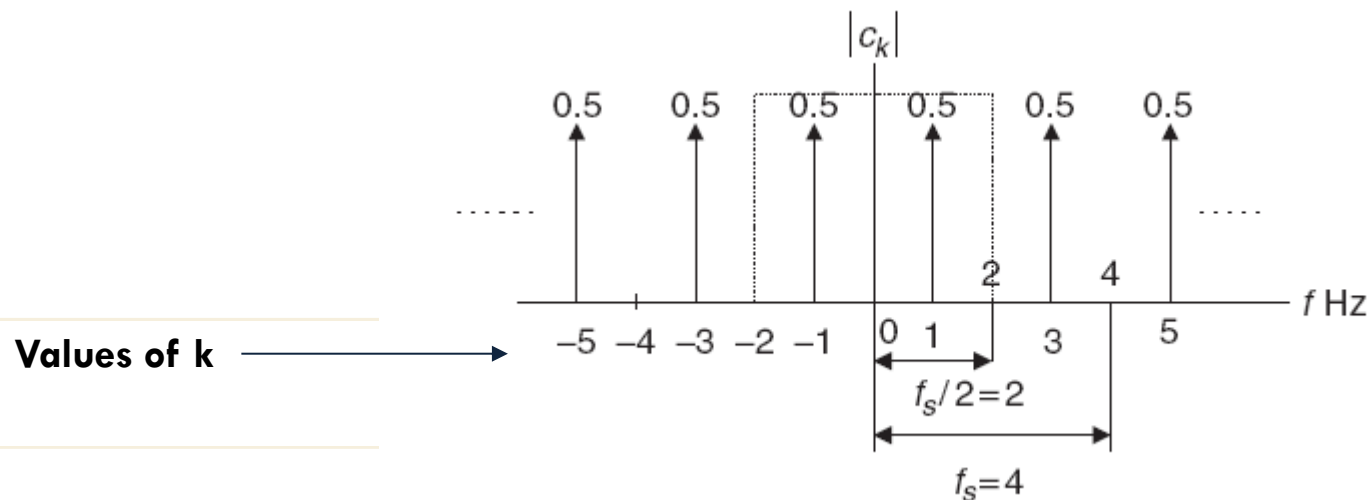
Similarly, after solving c_2 and c_3 , we'll get

$$c_2 = \frac{1}{4} \sum_{n=0}^3 x[n] e^{-j 2\pi \times \frac{2n}{4}} = 0$$

$$c_3 = \frac{1}{4} \sum_{n=0}^3 x[n] e^{-j 2\pi \times \frac{3n}{4}} = j0.5$$

Using Periodicity property $c_k = c_{k+N}$, we get

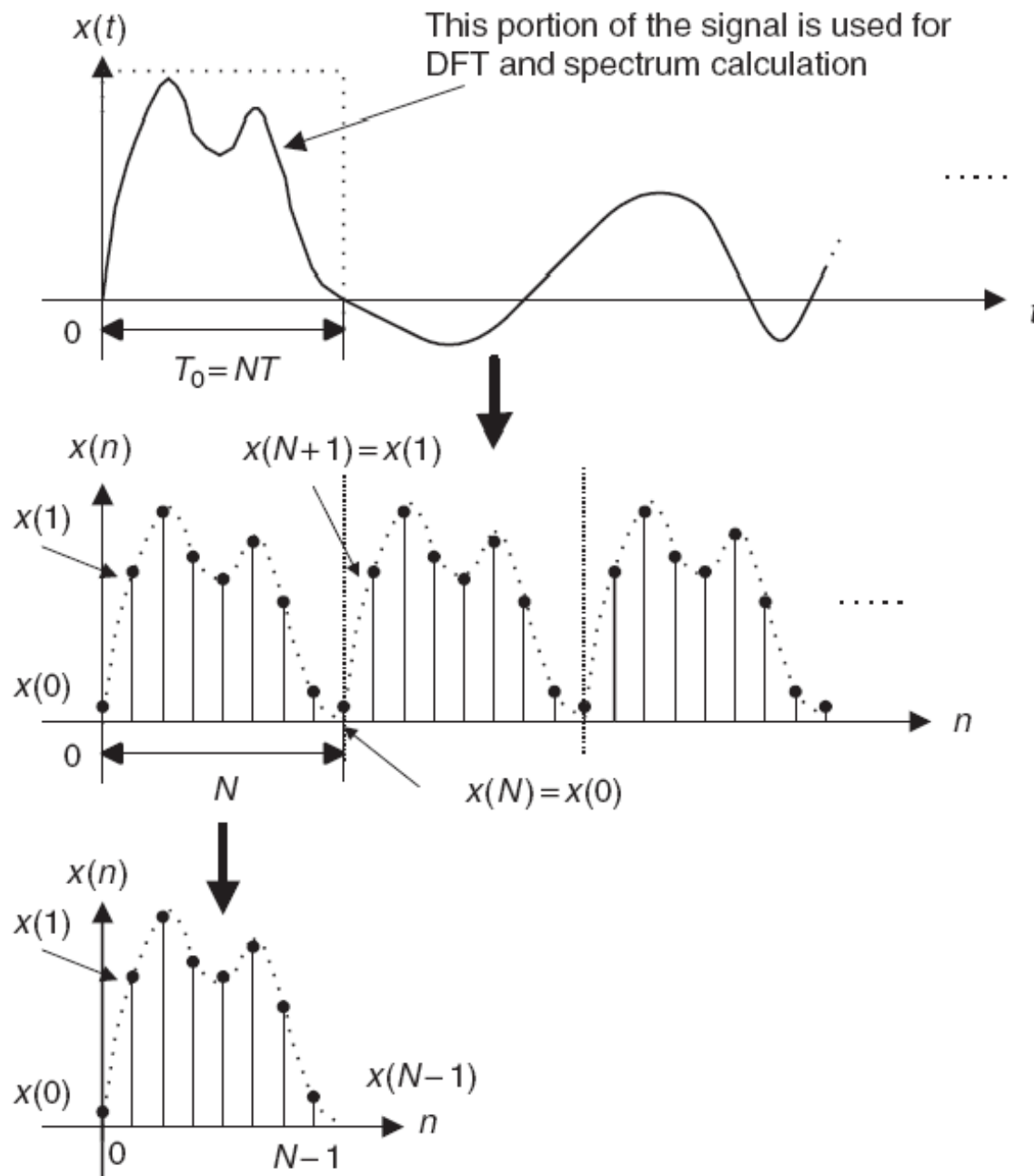
$$c_{-1} = c_3 = j0.5 \quad c_{-2} = c_2 = 0$$



Example 2: Given a sequence $x[n]$ for $0 \leq n \leq 3$ where $x[n] = [1 \ 2 \ 3 \ 4]$, find the Discrete Fourier Series

Discrete Fourier Transform

- Applies to periodic discrete-time sequences
 - ▣ We can also make an a-periodic sequence seem periodic
- Lets concentrate on development of DFT
 - ▣ First, we assume that the process acquires data samples from digitizing the interested continuous signal for a duration of T_0 seconds
 - ▣ Next, we assume that a periodic signal $x[n]$ is obtained by copying the acquired N data samples with the duration of T_0 to itself repetitively



□ Continued... developing DFT

- ▣ We determine the Fourier series coefficients using one-period N data samples (using equation for DFS)
- ▣ Multiply the coeffs by N to obtain i.e. $X(k) = Nc_k$

$$c_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}} \quad k = 0, 1, \dots, N-1 \quad \rightarrow \quad \text{DFS}$$

$$X(k) = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi kn}{N}} \quad k = 0, 1, \dots, N-1 \quad \rightarrow \quad \text{DFT}$$

where n is the time-index

and k is the frequency index, also known as 'bins'

Lets simply the equation by substituting,

$$W_N = e^{-j 2\pi / N} = \cos\left(\frac{2\pi}{N}\right) - j \sin\left(\frac{2\pi}{N}\right)$$

Therefore the Discrete Fourier Transform of $x[n]$ becomes

$$X(k) = \sum_{n=0}^{N-1} x[n] W_N^{kn} \quad k = 0, 1, \dots, N-1$$

and the Inverse Discrete Fourier Transform of $X[k]$ becomes

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X(k) W_N^{-kn} \quad n = 0, 1, \dots, N-1$$

Example 3: Given a sequence $x[n]$ for $0 \leq n \leq 3$ where $x[n] = [1 \ 2 \ 3 \ 4]$, find the Discrete Fourier Transform

Since $N = 4$, therefore $W_N^{kn} = W_4^{kn} = e^{-j2\pi kn/4} = e^{-j\pi kn/2}$

so now we can use the simplified formula

$$X(k) = \sum_{n=0}^3 x[n] W_4^{kn} = \sum_{n=0}^3 x[n] e^{-j\pi kn/2} \quad k = 0, 1, 2, 3$$

Solving for $k = 0$

$$\begin{aligned} X(0) &= \sum_{n=0}^3 x[n] e^{-j0} = x[0]e^{-j0} + x[1]e^{-j0} + x[2]e^{-j0} + x[3]e^{-j0} \\ &= 1 + 2 + 3 + 4 = 10 \end{aligned}$$

Solving for $k = 1$

$$\begin{aligned} X(1) &= \sum_{n=0}^3 x[n] e^{-j\frac{\pi n}{2}} = x[0]e^{-j0} + x[1]e^{-j\frac{\pi}{2}} + x[2]e^{-j\pi} + x[3]e^{-j\frac{3\pi}{2}} \\ &= x[0](1) + x[1](-j) + x[2](1) + x[3](-1) \\ &= 1 - j2 - 3 - j4 = -2 + j2 \end{aligned}$$

Similarly, solving for $k = 2$ and $k = 3$ we get

$$X(2) = -2$$

$$X(3) = -2 - j2$$

You can even verify this result yourself using Matlab, try this

```
fft([1 2 3 4])
```

```
ans = 10
```

```
-2+2i
```

```
-2
```

```
-2-2i
```

Example 4: Find the Inverse Discrete Fourier Transform of the DFT found in example 3

In example 2, we found out the following values in freq. domain

$$X(0) = 10 ; X(1) = -2+j2 ; X(2) = -2 ; X(3) = -2-j2$$

Since $N = 4$, $W_4^{-1} = e^{-\pi/2}$, therefore,

$$x[n] = \frac{1}{4} \sum_{k=0}^3 X(k) W_4^{-kn} = \frac{1}{4} \sum_{k=0}^3 X(k) e^{j\pi kn/2} \quad n = 0,1,2,3$$

Solving for $n = 0$

$$\begin{aligned} x[0] &= \frac{1}{4} \sum_{k=0}^3 X(k) e^{j0} = X(0)e^{j0} + X(1)e^{j0} + X(2)e^{j0} + X(3)e^{j0} \\ &= \frac{1}{4} \{10 + (-2 + j2) - 2 + (-2 - j2)\} = 1 \end{aligned}$$

Solving for $n = 1$

$$\begin{aligned}x[1] &= \frac{1}{4} \sum_{k=0}^3 X(k) e^{j\frac{\pi k}{2}} = X(0)e^{j0} + X(1)e^{j\frac{\pi}{2}} + X(2)e^{j\pi} + X(3)e^{j\frac{3\pi}{2}} \\&= \frac{1}{4} [X(0)\{1\} + X(1)\{j\} + X(2)\{-1\} + X(3)\{-j\}] \\&= \frac{1}{4} \{10 + j(-2 + j2) - (-2) - j(-2 - j2)\} = 2\end{aligned}$$

Similarly, solving for $n = 2$ and $n = 3$ we get

$$x[2] = 3$$

$$x[3] = 4$$

You can even verify this result yourself using Matlab, try this

```
ifft([10, -2+2i, -2, -2-2i])
```

```
ans = 1      2      3      4
```

Relationship between **k** and its associated frequency

The N-point DFT expresses frequency content from $-f_s/2$ Hz to $f_s/2$ Hz

So can we map the frequency bin 'k' to its corresponding frequency as

$$f = \frac{kf_s}{N} \quad \text{Hz}$$

The frequency resolution, the frequency step between two DFT points, is given as

$$\Delta f = \frac{f_s}{N} \quad \text{Hz}$$

Example 4: In example 3, we found the DFT of the given sequence $x[n]$

If the sampling rate is 10 Hz, determine

- (a) The sampling period, time index, and sampling time instant for a digital sample $x[3]$ in time domain
- (b) The frequency resolution, frequency bin number, and mapped frequency for each of the DFT coefficients $X(1)$ and $X(3)$ in frequency domain

(a) In time-domain, the sampling period can be calculated as

$$T = \frac{1}{f_s} = \frac{1}{10} = 0.1 \quad \text{sec}$$

For $x[3]$ the time index is $n = 3$ and the sampling time instant is

$$t = nT = 3 \times 0.1 = 0.3 \quad \text{sec}$$

(b) In frequency-domain, since the total number of DFT points is $N = 4$

$$\Delta f = \frac{f_s}{N} = \frac{10}{4} = 2.5$$

The frequency bin number for $X(1)$ is $k = 1$, so its associated freq is

$$\Delta f = \frac{kf_s}{N} = \frac{1 \times 10}{4} = 2.5 \quad \text{Hz}$$

Similarly, for $X(3)$ is $k = 3$, so its associated freq is

$$\Delta f = \frac{kf_s}{N} = \frac{3 \times 10}{4} = 7.5 \quad \text{Hz}$$

Amplitude & Phase response

Once we have $X(k)$, we can find its amplitude and phase response

$$X(k) = R(k) + jI(k)$$

$$|X(k)| = \sqrt{\{R(k)\}^2 + \{I(k)\}^2}$$

$$\varphi_k = \tan^{-1}\left(\frac{I(k)}{R(k)}\right)$$

Example 5: Plot the magnitude and phase response for the following discrete-time signal $x[n]=[1 \ 0 \ 01]$