

A Candidate to Replace PID Control: SISO Constrained LQ Control¹

James B. Rawlings
Department of Chemical Engineering
University of Wisconsin–Madison
Austin, Texas
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Outline

- Motivation. PID and Model-based control
- Proposal. CLQ control
- Examples and comparison
- Conclusions

Motivation: six myths of PID and LQ control

Myth 1: A PID controller is simpler to implement and tune than an LQ controller.■

Myth 2: A PID controller with model-based tuning is as good as model-based control for simple processes such as SISO, 1st order plus time delay.■

Myth 3: A well-tuned PID controller is more robust to plant/model mismatch than an LQ controller.■

Myth 3 (alternate version): LQ controllers are not very robust to plant/model mismatch■

Motivation: six myths of PID and LQ control

Myth 4: Integrating the tracking error as in PID control is necessary to remove steady-state offset. Applying some anti-windup strategy for this integrator is therefore necessary when an input saturates.■

Myth 5: For simple processes (SISO, 1st order plus time delay) in the presence of input saturation, a PID controller with a simple anti-windup strategy is as good as model predictive control.■

Myth 6: PID controllers are omnipresent because they work well on most processes.■

Introduction

- PID control for single-input single-output (SISO) systems shows up everywhere in chemical process applications and process control education.
- Tuning rules are presented in numerous texts and, surprisingly, remain a topic of current control research [1, 10].
- *Question:* is PID's popularity due to any concrete technological advantage?

Technical advantages ascribed to PID control

- PID is simple, fast, and easy to implement in hardware and software
- PID is easy to tune
- PID provides good nominal control performance
- PID is robust to model errors

Model-based Control

- Model-based control methods include linear quadratic (LQ) control of unconstrained systems, and model predictive control (MPC) of constrained systems
- MPC is regarded by many in process control as complex to implement and tune
- The robustness of LQ control to model error has been a topic of debate [2]
- Some claim that PID controllers outperforms MPC controllers in the rejection of unmeasured load disturbances [9]

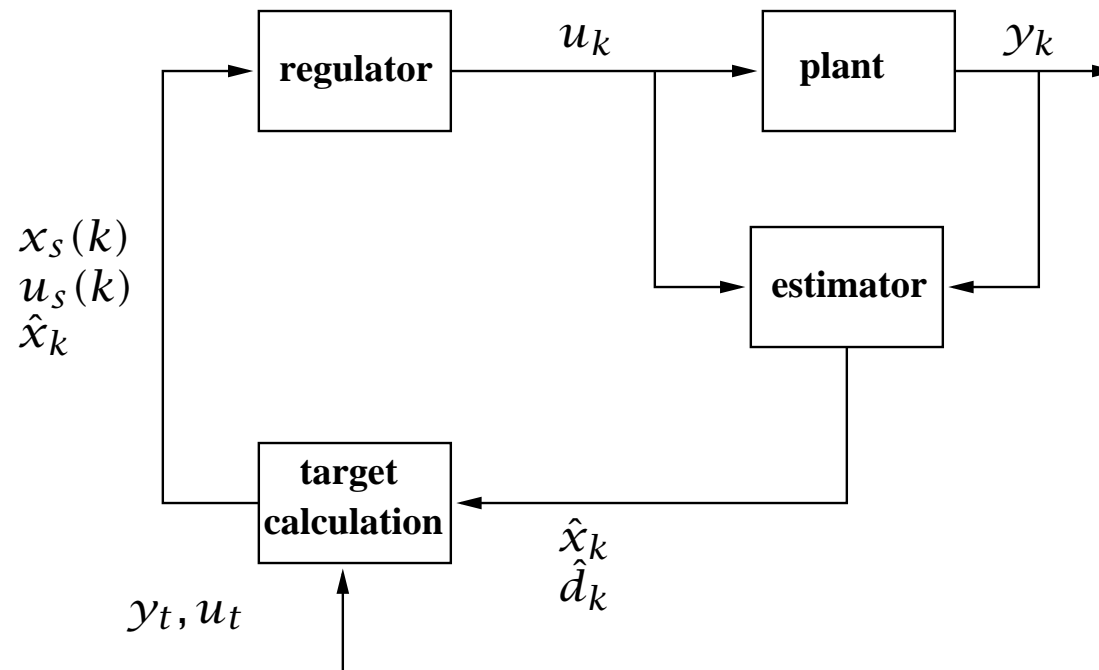
MPC's success in applications

- MPC has become the advanced controller of choice by industry mainly for the economically important, large-scale, multivariable processes in the plant
- The rationale for MPC in these applications is that the complexity of implementing MPC is justified only for the important loops with large payoffs

A modest proposal

- To address this perception of complexity, we propose a constrained, SISO linear quadratic controller (CLQ) with the following features:
- CLQ is essentially as fast to execute as PID (within a factor of five regardless of system order)
- CLQ is easy to implement in software and hardware
- CLQ displays both higher performance and better robustness than PID controllers

CLQ: regulation, estimation and steady-state targets



- Regulator: output/input penalty, Q/S .
- Estimator: disturbance variance/measurement variance, Q_d/R_v .

Implementation of CLQ

- Model: System *plus* disturbance to remove offset [4, 6, 5]

$$\begin{bmatrix} x \\ d \end{bmatrix}_{k+1} = \begin{bmatrix} A & B \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x \\ d \end{bmatrix}_k + \begin{bmatrix} B \\ 0 \end{bmatrix} u_{k-m}$$
$$y_k = \begin{bmatrix} C & 0 \end{bmatrix} \begin{bmatrix} x \\ d \end{bmatrix}_k$$

- Target: analytical solution for SISO case (fast)
- Regulator: the expensive part

- Estimator: unconstrained Kalman filter (fast)

$$\hat{x}_{k+1} = A\hat{x}_k + L_x(y_k - C\hat{x}_k)$$

$$\hat{d}_{k+1} = \hat{d}_k + L_d(y_k - C\hat{x}_k)$$

The regulator QP

Let $v = \{v_0, v_1, \dots, v_{N-1}\}$ be the sequence of inputs. We can write the regulator as a strictly convex QP:

$$\min_v \frac{1}{2} v^T H v + v^T c \quad (1a)$$

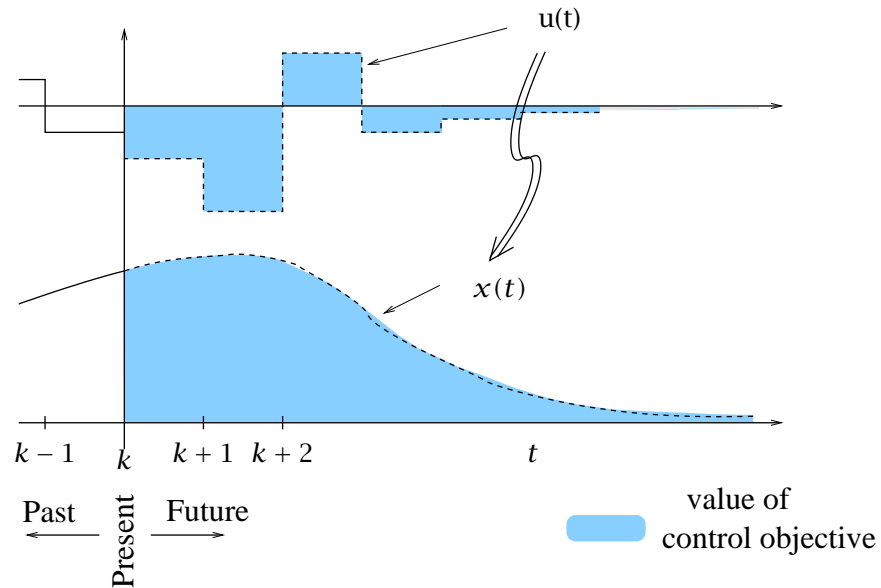
subject to:

$$\begin{bmatrix} \underline{u} \\ \underline{u} \\ \vdots \\ \underline{u} \end{bmatrix} \leq \begin{bmatrix} v_0 \\ v_1 \\ \vdots \\ v_{N-1} \end{bmatrix} \leq \begin{bmatrix} \overline{u} \\ \overline{u} \\ \vdots \\ \overline{u} \end{bmatrix} \quad (1b)$$

Let $v^* = (v_0^*, \dots, v_{N-1}^*)$ denote the optimal solution to (1). The current control input is

$$u_k = u_s(k) + v_0^*$$

Storing the active sets of the regulator



- Each control move u_k can be at the upper bound, at the lower bound, or somewhere in between.
- We construct all combinations of constraints, 3^N .
- For $N = 4$, $3^4 = 81$ different active sets.

Regulator implementation — Two basic steps

1. The offline generation of a solution table. This step involves solving linear equations, multiplications and additions.
2. The online table scanning given the current value of x . This step involves only multiplications and additions and checking conditionals. These same operations are required in PID control.

The active set table

$$u_0 = K_i x + b_i$$

$N = 2$ ■

i	constraint set	K_i	b_i
1	$\{\bar{u}, \bar{u}\}$	0	\bar{u}
2	$\{\bar{u}, -\}$	0	\bar{u}
3	$\{\bar{u}, \underline{u}\}$	0	\bar{u}
4	$\{-, \bar{u}\}$	K_4	b_4
5	$\{-, -\}$	K_5	b_5
6	$\{-, \underline{u}\}$	K_6	b_6
7	$\{\underline{u}, \bar{u}\}$	0	\underline{u}
8	$\{\underline{u}, -\}$	0	\underline{u}
9	$\{\underline{u}, \underline{u}\}$	0	\underline{u}

$N = 4$ ■

i	constraint set	K_i	b_i
1	$\{\bar{u}, \bar{u}, \bar{u}, \bar{u}\}$	0	\bar{u}
2	$\{\bar{u}, \bar{u}, \bar{u}, -\}$	0	\bar{u}
3	$\{\bar{u}, \bar{u}, \bar{u}, \underline{u}\}$	0	\bar{u}
...
40	$\{-, -, -, \bar{u}\}$	K_{40}	b_{40}
41	$\{-, -, -, -\}$	K_{41}	b_{41}
42	$\{-, -, -, \underline{u}\}$	K_{42}	b_{42}
...
79	$\{\underline{u}, \underline{u}, \underline{u}, \bar{u}\}$	0	\underline{u}
80	$\{\underline{u}, \underline{u}, \underline{u}, -\}$	0	\underline{u}
81	$\{\underline{u}, \underline{u}, \underline{u}, \underline{u}\}$	0	\underline{u}

Example 1 — First order plus time delay

- The first example is a first order plus time delay (FOPTD) system:

$$G_1(s) = \frac{e^{-2s}}{10s + 1} \quad \text{sampled with } T_s = 0.25$$

- The input is assumed to be constrained $|u| \leq 1.5$
- The control horizon is $N = 4$

Tuning

- The estimator is designed with $q_x = 0.05$ and $R_v = 0.01$ for both CLQ controllers
- The regulator input penalty is $s = 1$ for CLQ 1, and $s = 10$ for CLQ 2.
- The tuning parameters for PID 1 are chosen according to Luyben's rules [3, p. 97]: $K_c = 2.51$, $T_i = 17.3$, $T_d = 0$.
- The tuning parameters for PID 2 are chosen according to Skogestad's IMC rules [11]: $K_c = 2.35$, $T_i = 10$, $T_d = 0$.

Setpoint change and load disturbances

- In all simulations the setpoint is changed from 0 to 1 at time zero
- At time 25 a load disturbance passing through the same dynamics as the plant of magnitude -0.25 enters the system
- at time 50 the disturbance magnitude becomes -1 (which makes the setpoint 1 unreachable)
- finally at time 75 the disturbance magnitude becomes -0.25 again.

FOPTD system: nominal case.

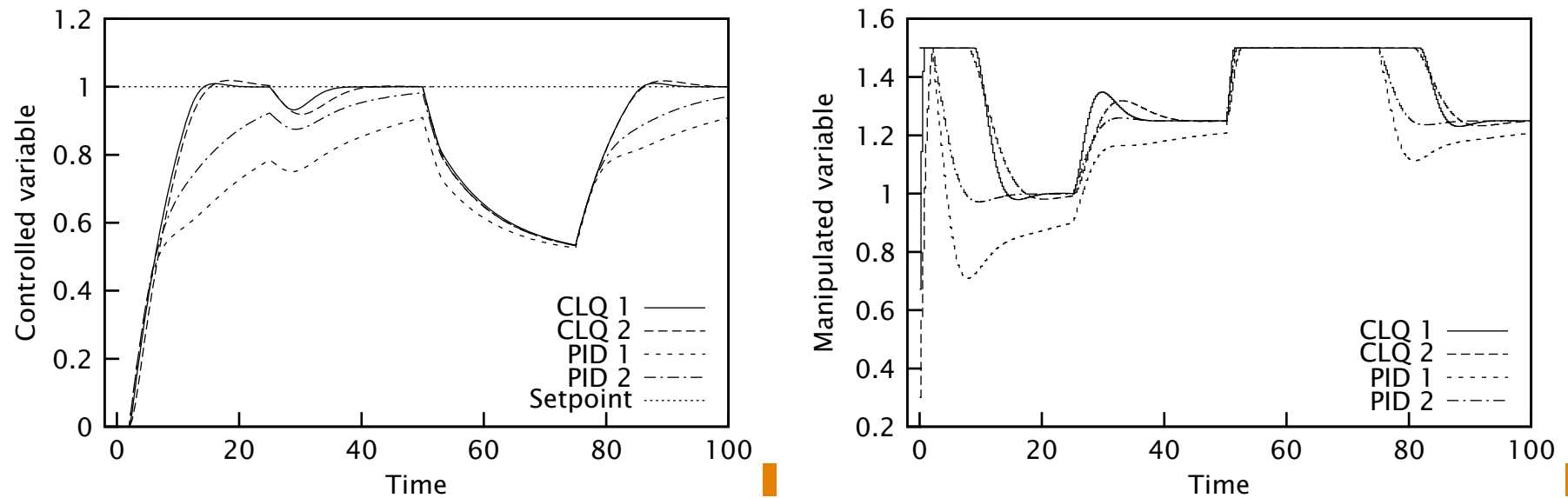


Figure 1: FOPTD system: nominal case.

FOPTD system: noisy case.

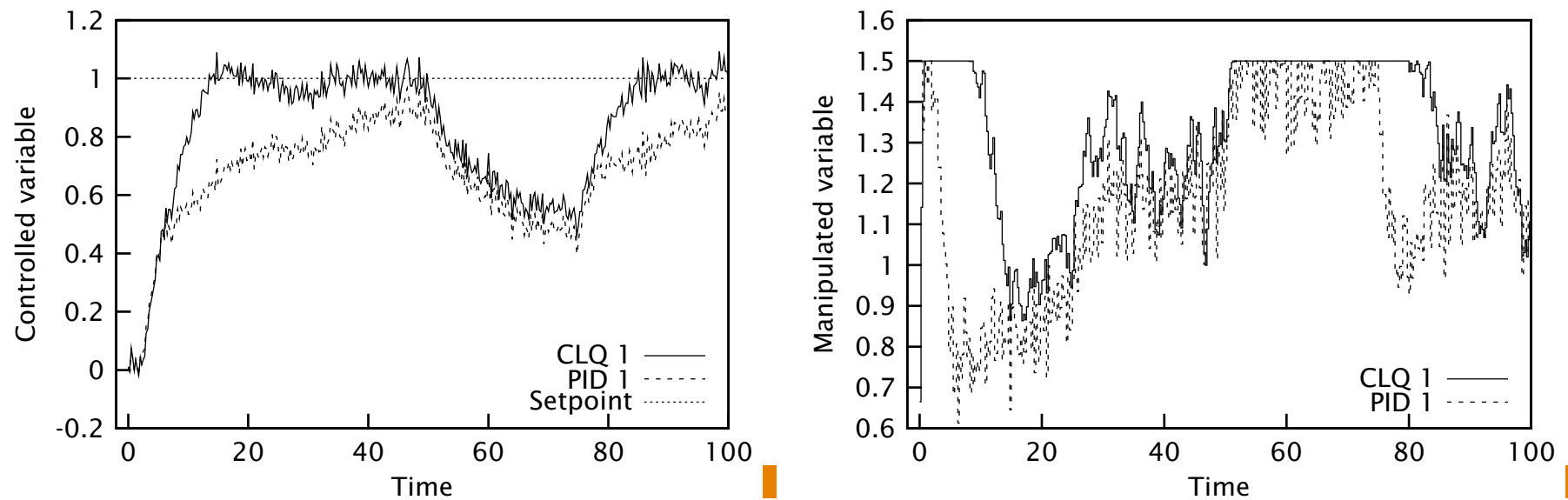


Figure 2: FOPTD system: noisy case.

FOPTD system: effect of plant/model mismatch.

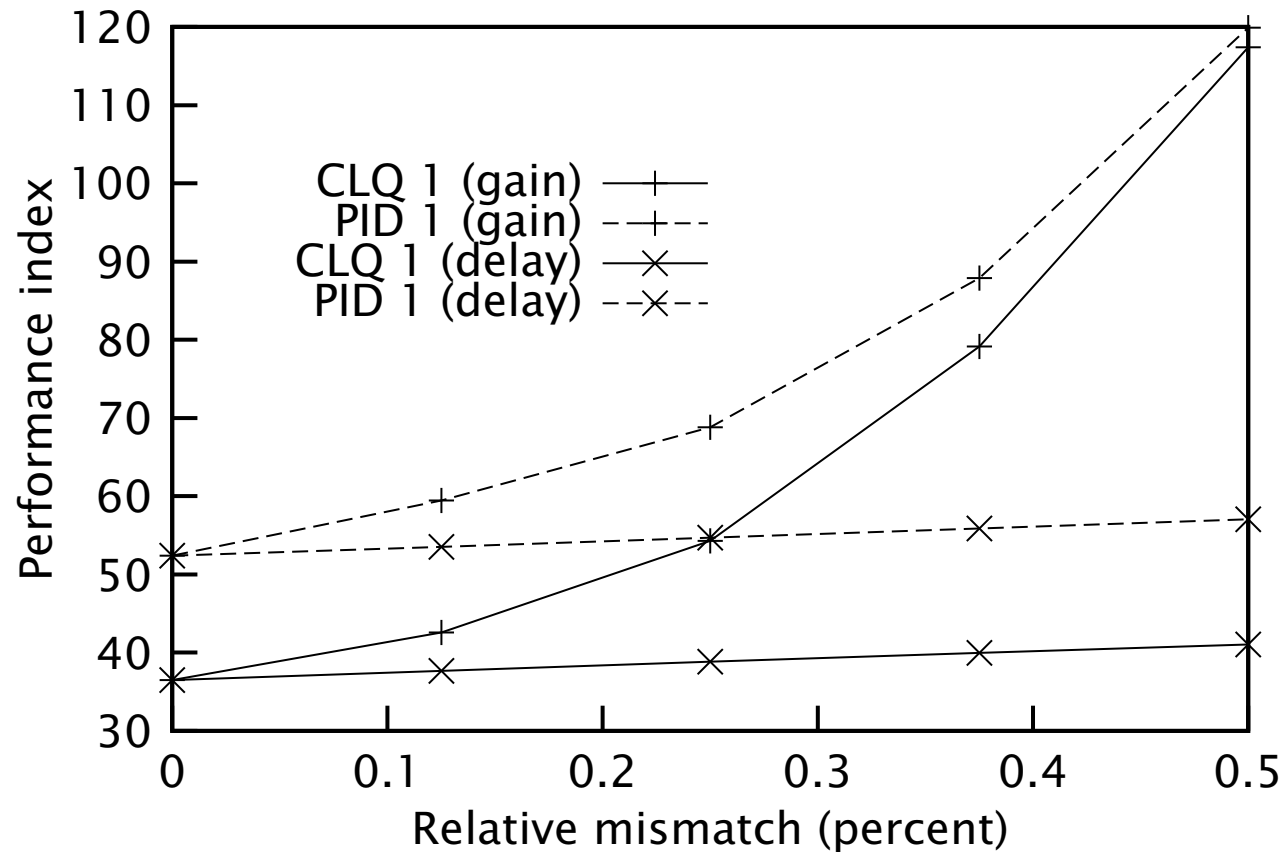


Figure 3: FOPTD system: effect of plant/model mismatch.

Example 2 — Integrating system

- The second example is an integrating system:

$$G_2(s) = \frac{e^{-2s}}{s} \quad \text{sampled with } T_s = 0.25$$

- The same input constraints, horizon, setpoint change and disturbances, and estimator parameters as in the first example are considered.
- CLQ 1 uses a regulator input penalty of $s = 500$, while CLQ 2 uses $s = 5000$.
- The tuning parameters for PID 1 are chosen according to Luyben's rules [3, p. 97]: $K_c = 0.23$, $T_i = 18.7$, $T_d = 0$.
- The tuning parameters for PID 2 are chosen according to Skogestad's IMC rules [11]: $K_c = 0.23$, $T_i = 17$, $T_d = 0$.

Integrating system: nominal case.

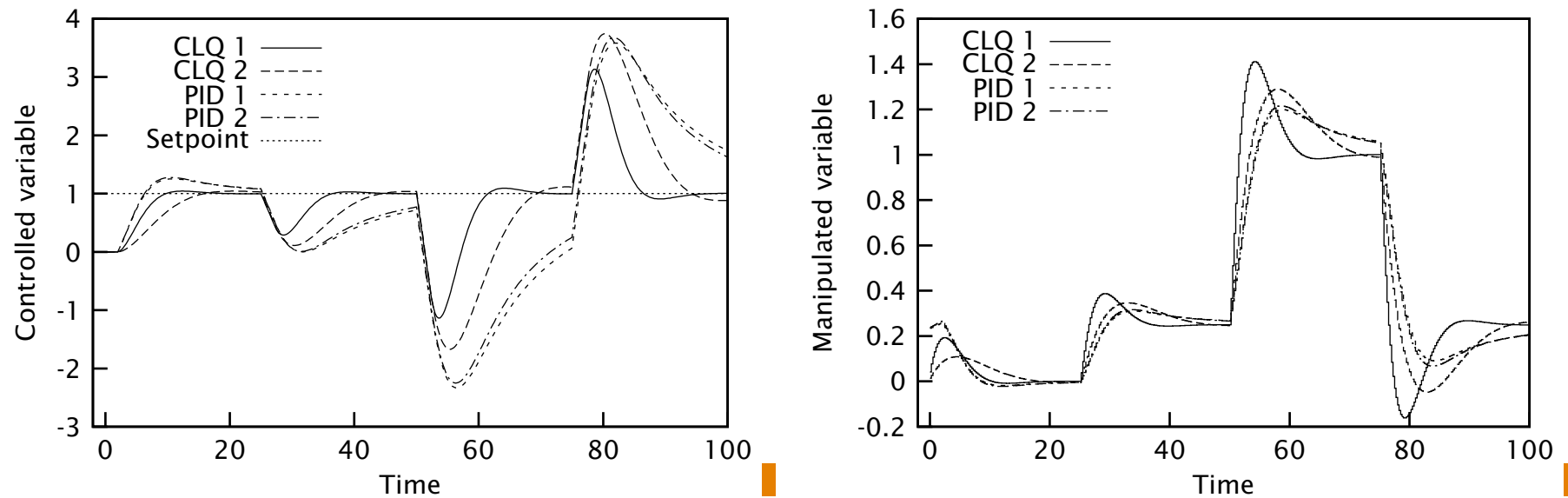


Figure 4: Integrating system: nominal case.

Integrating system: noisy case.

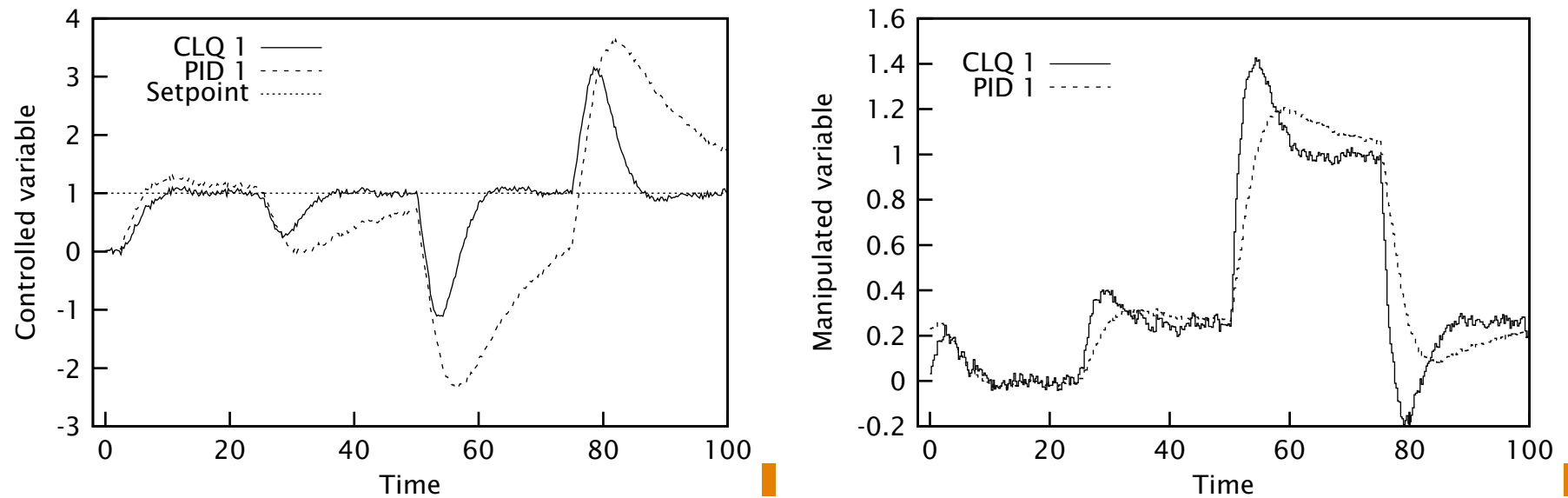


Figure 5: Integrating system: noisy case.

Integrating system: effect of plant/model mismatch.

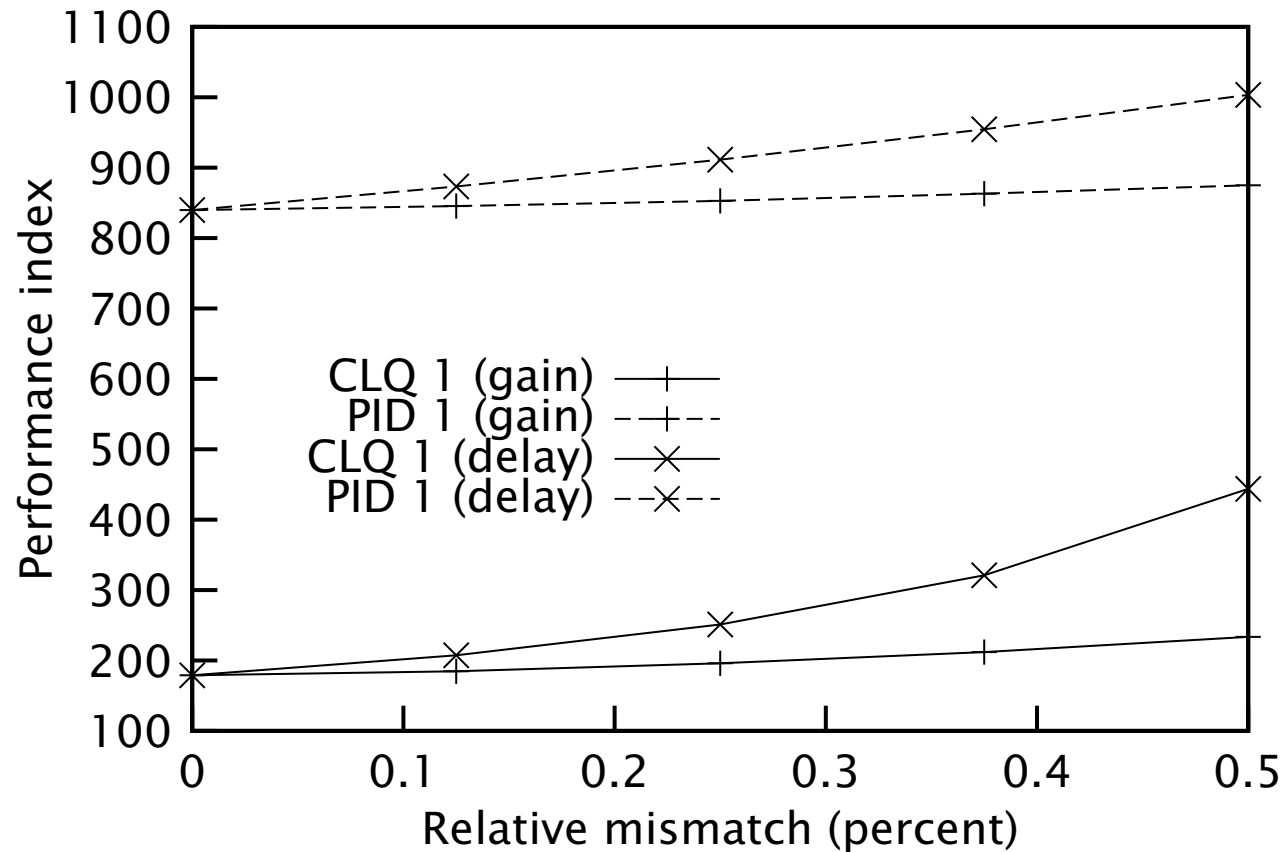


Figure 6: Integrating system: effect of plant/model mismatch.

Example 3 — Under-damped system

- The third example is a second-order, under-damped system:

$$G_3(s) = \frac{K}{\tau^2 s^2 + 2\tau\xi s + 1} \quad T_s = 0.25, K = 1, \tau = 5, \xi = 0.2$$

- The same input constraints, horizon, setpoint change and disturbances, and estimator parameters as in the first example are assumed.
- CLQ 1 uses a regulator input penalty of $s = 1$, while CLQ 2 uses $s = 10$.
- The tuning parameters for PID 1 are chosen according to Luyben's rules [3, p. 97]: $K_c = 7.29$, $T_i = 16.8$, $T_d = 1.21$.
- The tuning parameters for PID 2 are chosen following the same IMC approach as in [11]: $K_c = 0.40$, $T_i = 2$, $T_d = 12.5$.

Under-damped system: nominal case.

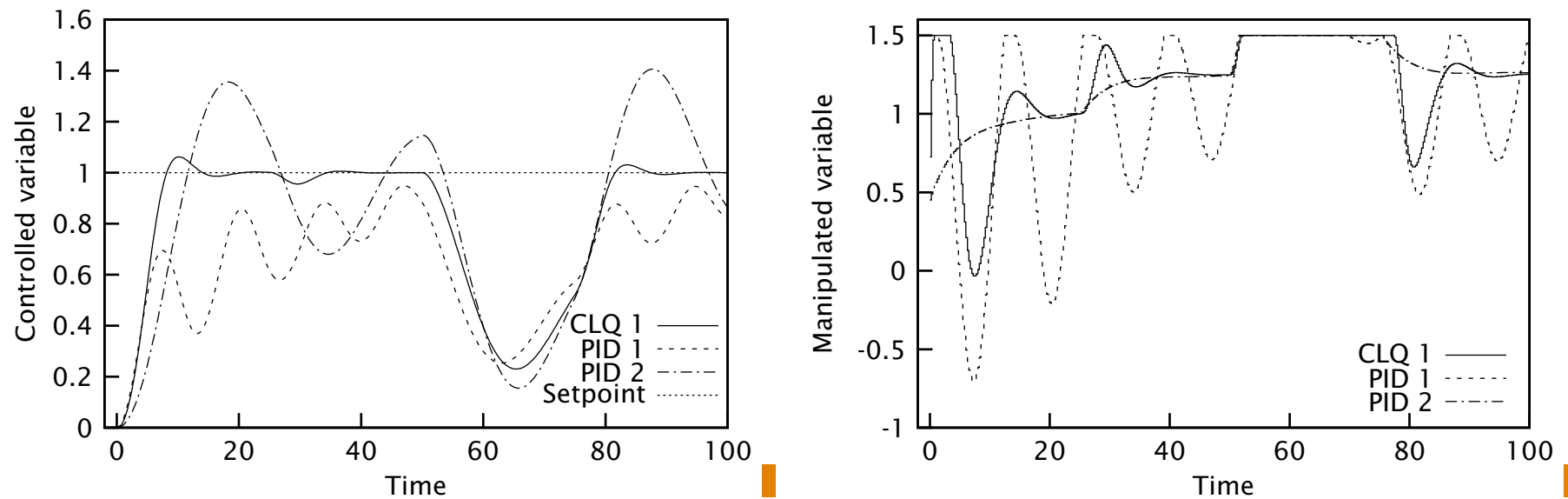


Figure 7: Under-damped system: nominal case.

Under-damped system: noisy case.

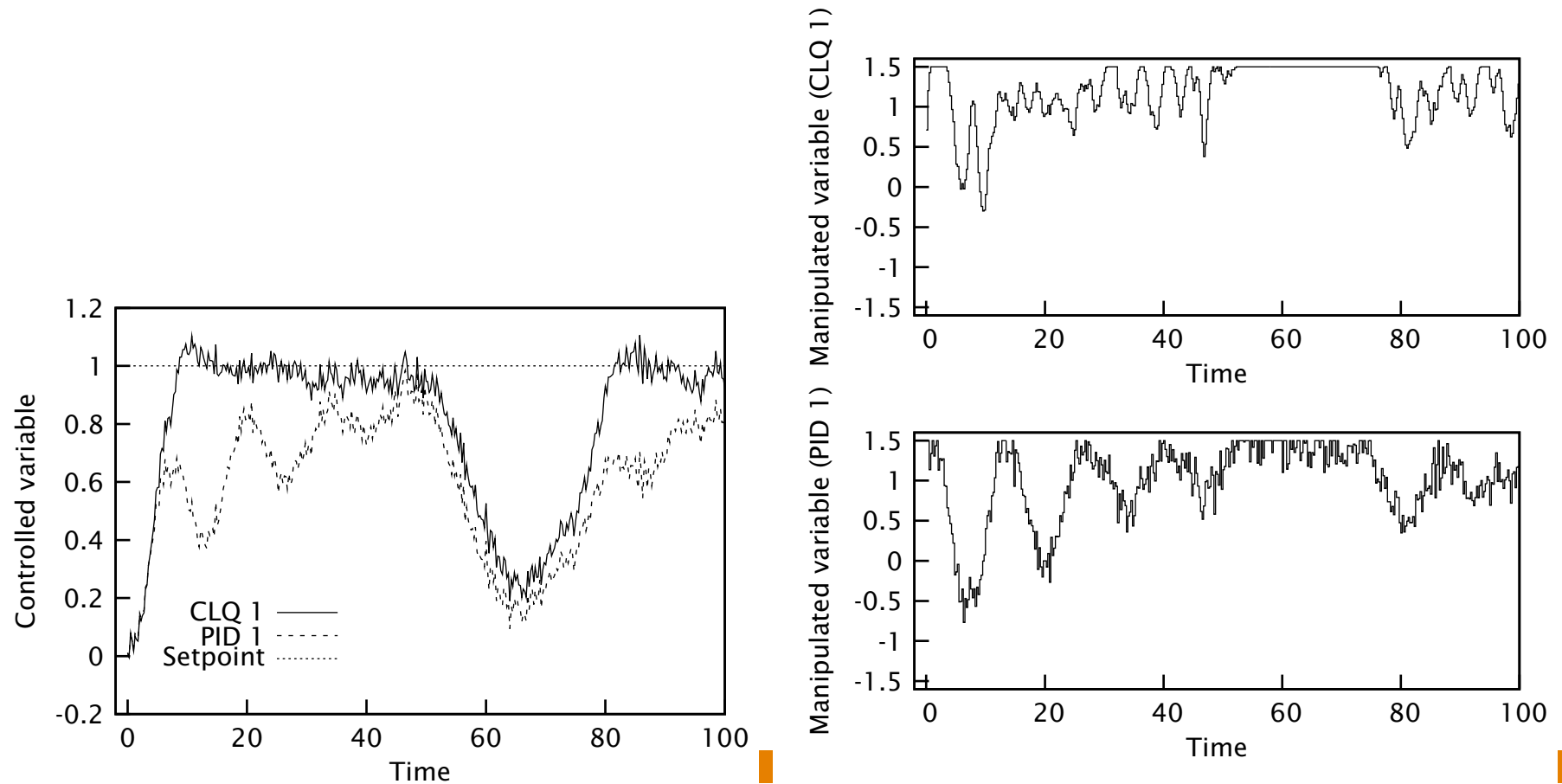


Figure 8: Under-damped system: noisy case.

Under-damped system: effect of plant/model mismatch.

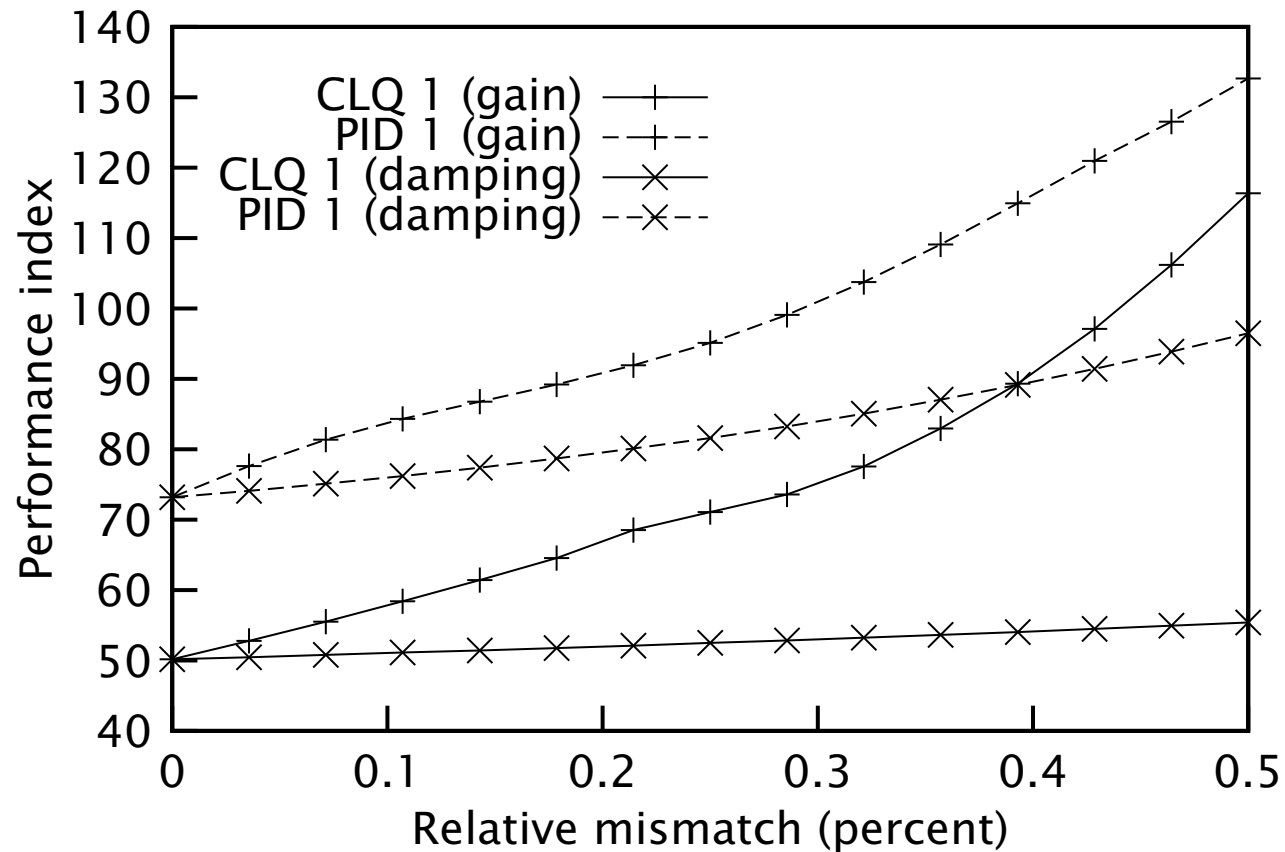


Figure 9: Under-damped system: effect of plant/model mismatch.

Computation time for CLQ

- The computational burden of CLQ is comparable to that of PID.

	average CPU time (ms)	maximum CPU time (ms)
PID	0.05	0.10
CLQ	0.22	0.55

- The CPU is a 1.7 GHz Athlon PC running Octave

Revisiting the six myths

Myth 1: A PID controller is simpler to implement and tune than an LQ controller.

- The validity of this myth rests largely with the hardware and control software vendors. Not difficult to implement CLQ if vendors offer on DCS.■
- Regarding tuning, it is not difficult to look up tuning rules for a PID controller.■
- It *is* difficult to find PID tuning parameters that give similar performance and robustness to an LQ controller.■
- The LQ controller is not difficult to tune. The effects of its two tuning parameters are clear.■

Myth 2: A PID controller with model-based tuning is as good as model-based control for simple processes such as SISO, 1st order plus time delay.

- No evidence to support this myth. Figure 1 shows the opposite is true.■
- If we restrict “simple process” to “first-order process,” this myth may remain in currency.■

Myth 3: A well-tuned PID controller is more robust to plant/model mismatch than an LQ controller.

- No evidence to support this myth. Figures 3, 6, and 9 show the opposite is true.■
- No superior robustness properties for PID control given any recommended tuning rules.■

Myth 3 (alternate version): LQ controllers are not very robust to plant/model mismatch.

- One can construct processes for which the state feedback regulator has good margins but output feedback with the same regulator and a state estimator has poor margins [2].■
- We have yet to see examples that indicate this issue has industrial significance.■

Myth 4: Integrating the tracking error as in PID control is necessary to remove steady-state offset. Applying some anti-windup strategy for this integrator is therefore necessary when an input saturates.

- Integrating the tracking error is not required for offset free control as shown in all of the examples■
- Integrating the model error is a sharper idea, and also removes the need for an anti-windup strategy when the input saturates.■

Myth 5: For simple processes (SISO, 1st order plus time delay) in the presence of input saturation, a PID controller with a simple anti-windup strategy is as good as model predictive control.

- The constraint handling properties of PID are not competitive with MPC.■
- Even for SISO, the difference can be noticeable. See Figures 1 and 7.■

Myth 6: PID controllers are omnipresent because they work well on most processes.

- Seeing no evidence that PID controllers work particularly well, consider an explanation rooted more in human behavior.■
- PID controllers are everywhere because vendors programmed them in the DCS when they replaced analog PID.■

Disruptive Technology — Benefits of CLQ

- Provides a single, scalable control technology ranging from the fastest SISO loop to the slowest, largest, MIMO dynamic plant optimization
- Because of the model forecast, constraints and optimization features, we can network many SISO CLQs together to achieve full benefits of multivariable MPC control
- Take advantage of these “smart controllers” embedded at all plant levels
- Modify model used in forecast as conditions change
- Rewrite objective function to achieve changing plant objectives

Disruptive Technology — Costs of CLQ

- Vendor companies will have to implement on the DCS
- Modest modeling cost (SISO step test)
- Operators will need new training
- Textbook materials will need to be revised
- Inertia will have to be overcome

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